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Matched Subspace Detectors for Stochastic Signals*

Todd McWhorter

Mission Research Corporation

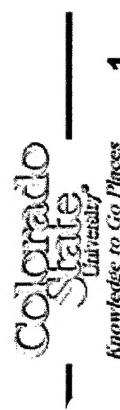
mcwhorter@aster.com

Louis Scharf

Colorado State University

scharf@enqr.colostate.edu

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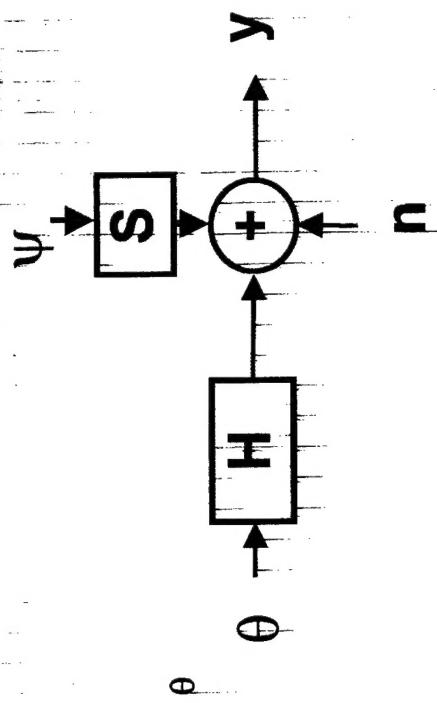
*Colorado
State
University*

Innovative to the Point

Problem Statement

- The goal is to design detectors for stochastic signals or second-order signals.

- Extension of the first-order matched subspace detectors of Scharf and Friedlander.



- It is assumed that interference is nulled prior to processing by projecting the data into the space orthogonal to the interference subspace.

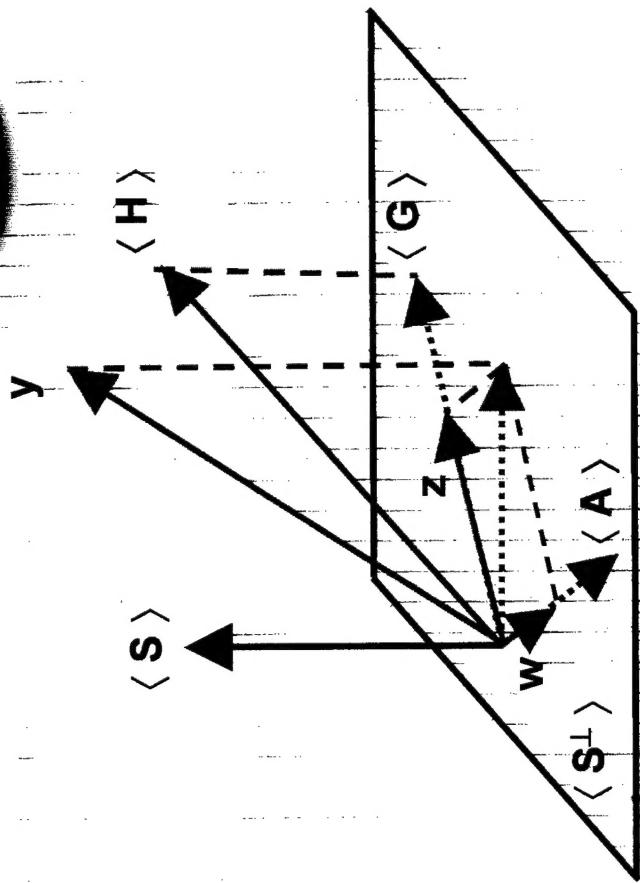
$$n \sim \mathcal{CN}(0, \sigma^2)$$

$$\theta \sim f(\theta; \beta) \text{ ex. } \theta \sim \mathcal{CN}(0, R_{\theta\theta})$$

- We assume various states of knowledge about the parameters σ^2 and β .

Pre-Processing

- In order to be invariant to the interference statistics, the data are projected into the space orthogonal to the interference.
- The data are then decomposed into their signal and noise components.



$$\begin{aligned} z &= (H^*(I-P_S)H)^{-1/2}H^*(I-P_S)y \\ &= (G^*G)^{-1/2}G^*y \end{aligned}$$

$$AA^* = I - P_S \cdot P_G \quad w = A^* y$$



Hypotheses

- The "noise" vector w is distributed as a white complex Gaussian vector regardless of which hypothesis is in effect.

$$\text{Define } \phi = (G^*G)^{1/2}.$$

- When signal is present the data vector z is distributed:

$$f(z | \phi) = \frac{1}{(\pi\sigma^2)^p} e^{-\frac{1}{\sigma^2} \|z - \phi\|^2}$$

- When signal is not present the data vector z is distributed:

$$f(z | \phi = 0) = \frac{1}{(\pi\sigma^2)^p} e^{-\frac{z^*z}{\sigma^2}}$$

Likelihood Ratio

- For now, assume that the noise power σ^2 is known.
- In this case the vector w is common to both hypotheses and is of no use.
- The *conditional likelihood ratio* is then

$$l(\mathbf{z} \mid \phi; \sigma^2) = \frac{f(\mathbf{z} \mid \phi; \sigma^2)}{f(\mathbf{z} \mid \phi=0; \sigma^2)}$$
$$= \exp\left(\frac{\mathbf{z}^* \mathbf{z}}{\sigma^2}\right) \times \exp\left(-\frac{1}{\sigma^2} \|\mathbf{z} - \phi\|^2\right)$$

Unconditional Likelihood Ratio

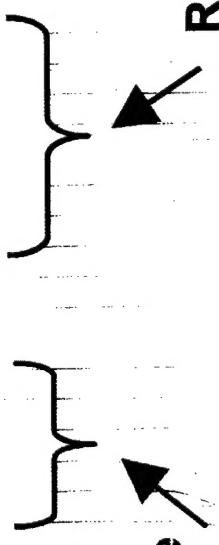
- The unconditional likelihood ratio can be written as

$$l(z; \sigma^2, \beta) = \exp\left(\frac{z^* z}{\sigma^2}\right) \times \int \exp\left(-\frac{\|z - \phi\|^2}{\sigma^2}\right) f_\phi(\phi; \beta) d\phi$$

- The log-likelihood ratio becomes

$$s(z; \sigma^2, \beta) = \frac{z^* z}{\sigma^2} + \ln \int \exp\left(-\frac{\|z - \phi\|^2}{\sigma^2}\right) f_\phi(\phi; \beta) d\phi$$

$$= \frac{y^* P_{Gy}}{\sigma^2} - p_r(z; \sigma, \beta)$$



Matched Subspace
Detector



Resolution Penalty

- The resolution penalty occurs because we presume to know something about the coordinate vector θ .
- If z is far from the “favored” orientation defined by θ then the penalty is larger than if the converse were true.

$$p_r(z; \sigma^2, \beta) = \ln \int \exp\left(-\frac{\|z - \phi\|^2}{\sigma^2}\right) f_\phi(\phi; \beta) d\phi$$

Gaussian Coordinate Vectors

- Suppose $\phi \sim \mathbf{CN}(0, R_{\phi\phi})$.
- Write the eigenvalue decomposition of $R_{\phi\phi}$ as:

$$R_{\phi\phi} = (G^*G)^{1/2} R_{\theta\theta} (G^*G)^{1/2} = V D^2 V^*$$

$V = [v_1 \ v_2 \ \cdots \ v_p]$; unitary

$$D^2 = \text{diag}[\beta_1^2, \beta_2^2, \cdots, \beta_p^2]$$

- Define the resolved signal-plus-noise to noise ratios:

$$r_i = 1 + \frac{\beta_i^2}{\sigma^2}$$

Gaussian Penalty Term

- After some algebra the penalty term can now be written as

$$\begin{aligned} p_r(z, \sigma^2, \beta^2) &= -\ln \int \exp\left(-\frac{\|z-\phi\|^2}{\sigma^2}\right) \frac{1}{\pi^p \det(R_{\phi\phi})} \exp(-\phi^* R_{\phi\phi}^{-1} \phi) d\phi \\ &= \sum_{i=1}^p \ln(r_i) + \sum_{i=1}^p \frac{(z^* P_{V_i} z / \sigma^2)}{r_i} \end{aligned}$$

- This result implies that if the estimated signal-plus-noise to noise ratio $(z^* P_{V_i} z / \sigma^2)$ in the resolved subspace defined by V_i greatly exceeds r_i , then the penalty is large because of this mismatch.



Unknown Signal Power and Orientation



- Suppose that when signal is present we do not know $R_{\phi\phi}$.

- Recall the penalty term is

$$p_r(z; \varphi^2, \beta^2) = -\ln \int \exp\left(-\frac{\|z - \phi\|^2}{\beta^2}\right) \frac{1}{\pi^p \det(R_{\phi\phi})} \exp(-\phi^* R_{\phi\phi}^{-1} \phi) d\phi$$

$$= \sum_{i=1}^p \ln(r_i) + \sum_{i=1}^p \frac{(z^* P_{V_i} z / \sigma^2)}{r_i}$$

- The estimates of the signal-plus-noise to noise ratios are

$$r_i = \max(1, z^* P_{V_i} z / \sigma^2)$$

- We assume that $r_i \geq 1$ in the sequel.

Estimating Orientation

- The estimates of r_i in the previous slide depend on the orientation of the vectors v_i .

- We want to minimize

$$\prod_{i=1}^p \frac{z^* P_{v_i} z}{\sigma_i^2}$$

- We must also satisfy the constraints

$$\sum_{i=1}^p \frac{z^* P_{v_i} z}{\sigma_i^2} = \frac{z^* z}{\sigma^2}$$

$$r_i = \frac{z^* P_{v_i} z}{\sigma_i^2} \geq 1$$

Intermediate Orientation Solution

- The solution to this optimization problem is

$$r_i = \frac{\mathbf{z}^* P_{V_i} \mathbf{z}}{\sigma^2} = 1 \quad \text{for } i = 1, 2, \dots, p-1$$

$$r_p = \frac{\mathbf{z}^* P_{V_p} \mathbf{z}}{\sigma^2} = \frac{\mathbf{z}^* \mathbf{z}}{\sigma^2} - (p-1)$$

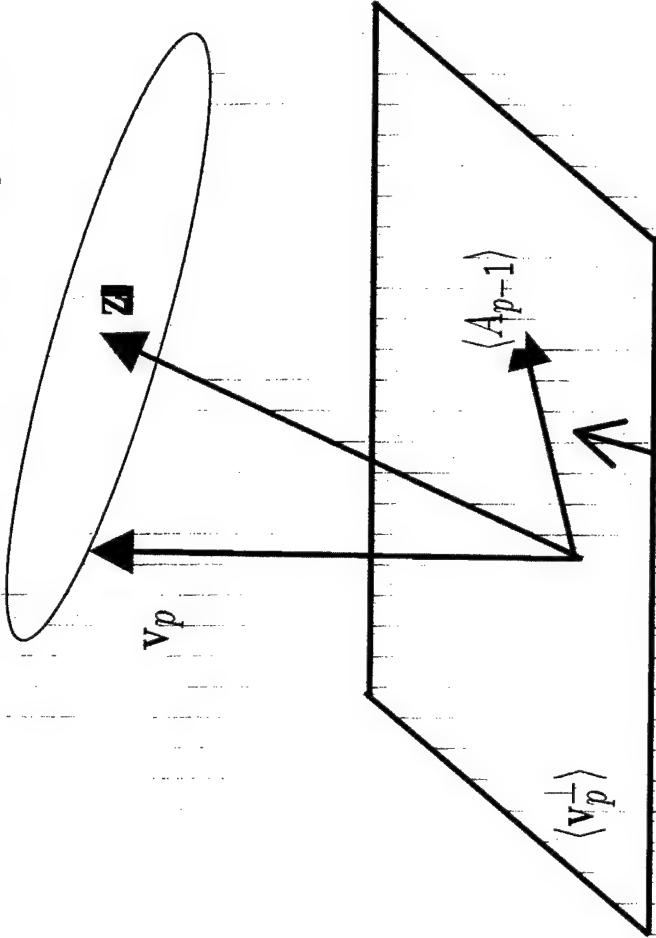
- The question remains: Is there a decomposition of $\langle G \rangle$ that has the above properties?
- The answer is yes.

Orientation Solution

- Solve for v_p first.
- Choose a v_p on the spherical invariance set defined by

$$\frac{z^* P_{V_H} z}{\sigma^2} = \frac{z^* z}{\sigma^2} - (p-1).$$

Great circle on invariance sphere



- Repeat this procedure in the spaces $\langle A_{p-1} \rangle, \langle A_{p-2} \rangle, \dots, \langle A_1 \rangle$

Has norm
 $\sigma^{2(p-1)}$



Compressed Likelihood

- Compressing the likelihood ratio with this solution gives the statistic

$$s(z, \sigma^2, \hat{R}_{\phi\phi}) = \frac{y^* P_{HY}}{\sigma^2} + \left[\ln\left(\frac{y^* P_{HY}}{\sigma^2}\right) - \text{constants} \right]$$

- This statistic is a monotonic function of the matched subspace detector. We can therefore use the MSD as the detection statistic

$$s = \frac{y^* P_{HY}}{\sigma^2}$$

- Then the result for 2nd-order models is the same as for 1st-order models.

Unknown Noise Power

- In the case of unknown noise power the GLRT detector can be written as a sum of the CFAR matched subspace detector and a penalty term

$$s(z; \hat{\sigma}^2, \hat{R}_{\phi\phi}) = \ln(1 + \hat{s}) + [\ln(\hat{s}) - \text{constants}]$$

- We can equivalently use the statistic

$$\hat{s} = \frac{\mathbf{y}^* P_H \mathbf{y}}{\hat{\sigma}^2}, \quad \hat{\sigma}^2 = \frac{1}{N-p} \mathbf{y}^* (I - P_H) \mathbf{y}$$

- These detectors are identical to the 1st-order results.

Rank-One Assumptions

- Here we assume that the signal subspace is rank-one.
- The complex-valued signal amplitude is written in polar form

$$\theta = M e^{j\phi}$$

- Assume that the phase and magnitude are uncorrelated and that the phase is uniformly distributed over $[0, 2\pi]$.

- Assume that the signal magnitude has a generalized Rayleigh distribution

$$f_M(M) = \frac{2M}{\beta^2} \left(\frac{M^2}{\beta^2}\right)^L \frac{e^{-M^2/\beta^2}}{L!}$$



Detectors with Known Noise

- $L=0$. This is the previous results with complex Gaussian amplitudes.

- $L \neq 0$. The penalty function is

$$p_r = (L+1) \ln(r) + \frac{(y^* P_h y / \sigma^2)}{r} \ln \left[\sum_{k=0}^L \frac{\text{binom}(L, k) \left((y^* P_h y / \sigma^2) (r-1) \right)^k}{k!} \right]$$

- Minimize the penalty term with respect to $r = 1 + \beta^2 / \sigma^2$.
- Compress the likelihood function with this term to obtain

$$s_r = \frac{y^* P_h y}{\sigma^2} + p_r(\hat{r}).$$